

# Accelerator Magnet Design

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# References - Acknowledgments



- USPAS course "Superconducting Accelerator Magnets", Ezio Todesco, Paolo Ferracin, Soren Prestemon
- USPAS Course "Magnetic Systems: Insertion Devices", Ross Schlueter

### Outline

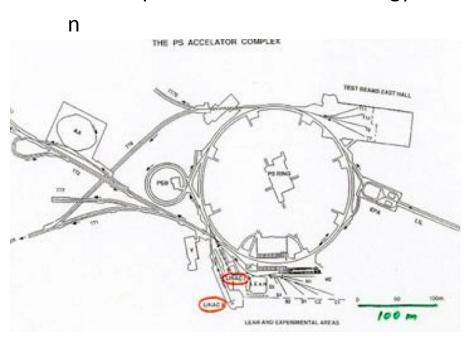


- The magnets of an accelerator
- Some magnetics fundamentals
- Review of magnetic multipoles
  - Definition: Taylor series
  - Inverse problem: how to create multipole fields
    - Iron-dominated (scalar potential)
    - Biot-Savart
- Design and fabrication issues with real accelerator magnets

### Layout of an accelerator



- Magnets play key role:
  - Kick beam into accelerator during injection: Kicker magnets
  - Align injected beam with stored beam: Septum magnets
  - Bend beam in circle: bend magnets (dipoles)
  - Focus beam to allow storage (quadrupoles)
  - Compensate for electron energy variation (sextupoles)





## Additional magnet systems



#### Correctors

- Dipoles for field trajectory correction
- Can be "slow": compensate static or slow-varying drifts
- Can be fast: allow fast feedback for beam control

#### Chicanes

- Versions of corrector magnets (not used for beam feedback)
- Used to provide mild steering of beam, e.g. in straights, or for dispersion
- Light-source Wigglers and undulators
  - Used in to produce synchrotron radiation of particular quality
  - Ideally are transparent to beam storage

# **Examples of Accelerator Magnets**



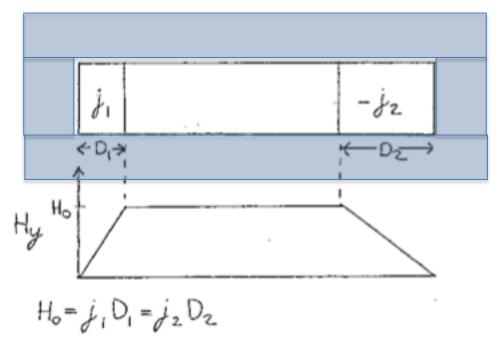
- Dipoles Steering
- Quadrupoles focusing
- Sextupoles chromaticity

- These components are analogous to optical elements, e.g. mirrors
  - Charged-particle optics

## Simplest dipole: windowframe



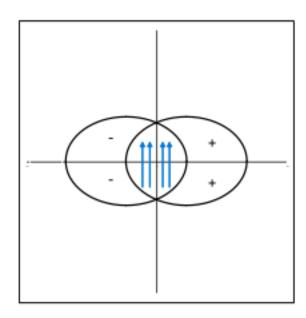
- Assume μ=∞; field in center is uniform
- Field across coil is linear
- What happens if μ is finite?
- What happens at the ends?



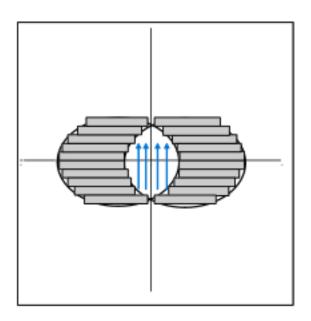
# Another "simple" geometry



- Constant J in ellipse => J=0 in intersecting zone
  - Field in center is uniform => perfect dipole
  - This is the motivation for standard " $Cos(\theta)$ " dipoles



Intersecting ellipses

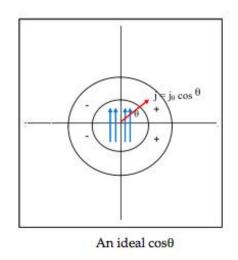


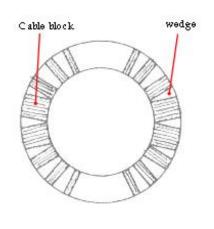
A practical (?) winding with flat cables

### Transitioning from theory to practice

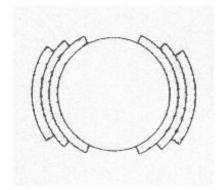


- Coil is made from a wire/cable => J~constant
  - Discretize  $Cos(\theta)$  distribution using wedges
  - Ends must allow beam-passage
  - These "details" introduce errors in the form of harmonic content

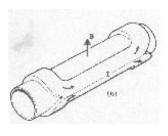




A practical winding with one layer and wedges [from M. N. Wilson, pg. 33]



A practical winding with three layers and no wedges [from M. N. Wilson, pg. 33]

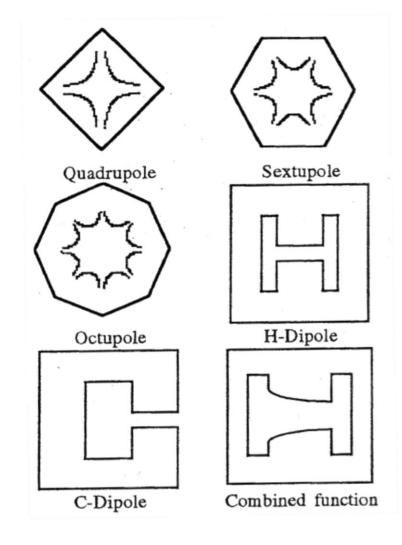


Artist view of a cosθ magnet [from Schmuser]

## Some classic configurations



- These configurations are often mentioned in the literature
  - Combined-function magnets can take a variety of forms
    - Scalar potential can define combination of fields
    - Scalar potential can be defined for "dominant" multipole of interest – other multipoles are then added via additional energization



### Review: Maxwell



Ampere

$$\nabla \times \vec{H} = \vec{J}$$

Faraday

$$\nabla \times \vec{E} = \vec{B}$$

$$\nabla \cdot \vec{B} = 0$$

$$\oint \vec{H} \cdot d\vec{s} = \int \vec{J} \cdot d\vec{a}$$

$$\oint \vec{E} \cdot d\vec{s} = -\int \dot{\vec{B}} \cdot d\vec{a}$$

No magnetic monopoles

$$\vec{B} = \mu \mu_0 \vec{H}, \quad \mu_0 = 4\pi 10^{-7} \frac{\text{V s}}{\text{A m}}$$

Continuity across interfaces implies:



$$\nabla \cdot \vec{B} = 0 \implies \Delta B_{\perp} = 0$$



$$\nabla \times \vec{H} = 0 \implies \Delta H_{||} = 0$$

# Allowed multipoles



$$F \equiv A + iV \tag{3}$$

It follows directly from Eqns. (2) and (3) that the complex conjugate  $B^*(z)$  of the field is analytic in z and is given by:

$$B^{*}(z) = i\frac{dF}{dz} \tag{4}$$

It is convenient to expand the complex  $\mathcal{K}_{\mathcal{A}}$  potential F in a power series about a point (say z=0) and analyze the 'harmonic' components:

$$F(z) = \sum_{n=1}^{\infty} \left(\frac{z}{r_p}\right)^n c_n \; ; \qquad B^{\bullet}(z) = i \sum_{n=1}^{\infty} \left(\frac{z}{r_p}\right)^{n-1} \frac{nc_n}{r_p} \tag{5}$$

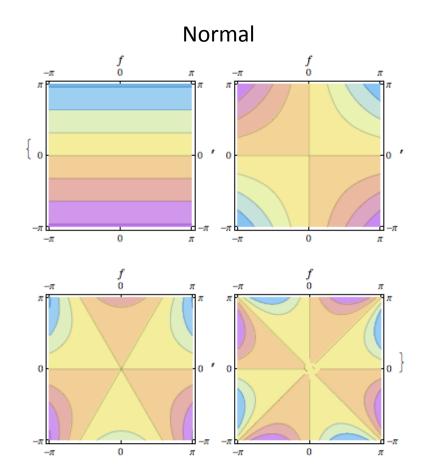
where  $r_p$  is the magnet aperture radius ( $\longrightarrow$  half gap h for a dipole). For magnets exhibiting midplane symmetry, the coefficients  $c_n \equiv a_n + ib_n$  are pure real (or pure imaginary if A, rather than V, is constant along the midplane). For symmetric multipole magnets (i.e. rotatable by  $360^o/2m$  with a change of polarity), of order m (e.g. m=1 for dipole, 2 for quadrupole, etc.) the complex potential F and flux density  $B^*(z)$  are

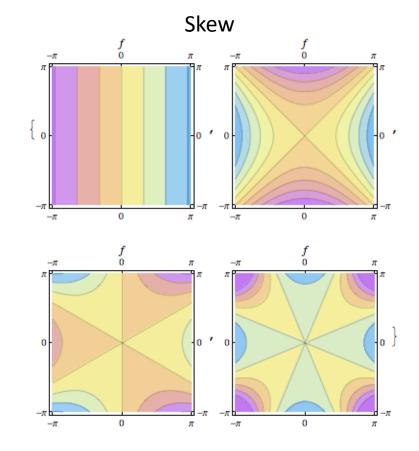
$$F(z) = \sum_{n=1}^{\infty} \left(\frac{z}{r_p}\right)^{m(2n-1)} a_{m(2n-1)}; \quad B^{\bullet}(z) = i \sum_{n=1}^{\infty} \left(\frac{z}{r_p}\right)^{m(2n-1)-1} \frac{m(2n-1)a_{m(2n-1)}}{r_p}$$
(6)

# Multipole fields



• Potential isosurfaces, m=1, 2, 3, 4

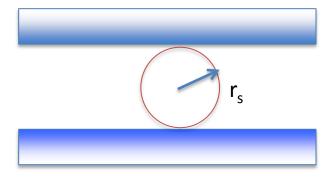




### Some comments



• The series expansion is only valid out to the minimum radius  $r_s$  of any potential surface



- Non-dimensionalization by  $r_s$  is often replaced by  $R_{ref}$ , a convenient measurement radius
- The coefficients beyond  $B_m$  (the dominant mode) are then often normalized by  $10^{-4}B_m$ 
  - resulting terms are said to be in "units"

#### FIELD HARMONICS OF A CURRENT LINE



Field given by a current line (Biot-Savart law)

$$B^*(z) = \frac{\mu_0 I}{2\pi i} \frac{1}{z - z_0}$$

$$\Longrightarrow F(z) = -\frac{\mu_0 I}{2\pi} Ln(z - z_0)$$

Or, in terms of multipoles:

$$F(z) = -\frac{\mu_0 I}{2\pi} \sum_{n=1}^{\infty} \frac{1}{n} \left(\frac{z}{z_0}\right)^n$$

$$= -\frac{\mu_0 I}{2\pi} \sum_{n=1}^{\infty} \frac{1}{n} \left(\frac{R_{ref}}{z_0}\right)^n \left(\frac{z}{R_{ref}}\right)^n$$



Félix Savart, French (June 30, 1791-March 16, 1841)



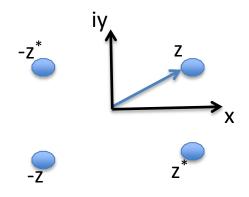
Jean-Baptiste Biot, French (April 21, 1774 – February 3, 1862)

#### FIELD HARMONICS OF A CURRENT LINE

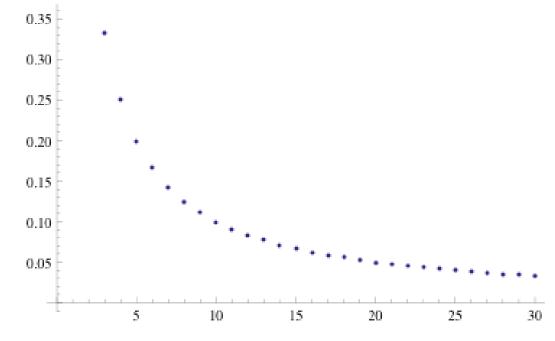


### The multipoles of a line current then scale like 1/n

- The details of the decay depend on the line current position
- Adding multiple line currents judiciously positioned can result in a multipole field of order m with fairly small multipoles n≠m



The line currents can be connected so as to create a dipole, quadrupole, etc



#### HOW TO GENERATE A PERFECT FIELD



#### Perfect dipoles

Cos theta: proof – homework from last Monday

$$j(\theta) = j_0 \cos(m\theta)$$

The vector potential reads

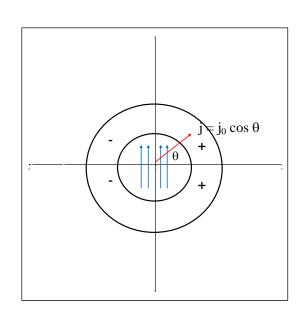
$$A_{z}(\rho,\phi) = \frac{\mu_{0} j}{2\pi} \sum_{n=1}^{\infty} \frac{1}{n} \left(\frac{\rho}{\rho_{0}}\right)^{n} \cos[n(\phi - \theta)]$$

and substituting one has

$$A_z(\rho,\phi) = \frac{\mu_0 j_0}{2\pi} \sum_{n=1}^{\infty} \frac{1}{n} \left(\frac{\rho}{\rho_0}\right)^n \int_0^{2\pi} \cos(m\theta) \cos[n(\phi-\theta)] d\theta$$

using the orthogonality of Fourier series

$$A_z(\rho,\phi) = \frac{\mu_0 j_0}{2m} \left(\frac{\rho}{\rho_0}\right)^m \cos(m\theta)$$



# Basic features of "sector" coils (Ezio Todesco)

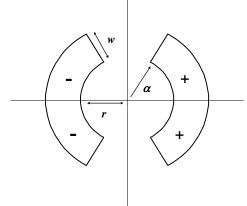


 We compute the central field given by a sector dipole with uniform current density i

$$I \rightarrow j\rho d\rho d\theta$$

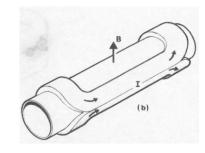
Taking into account of current signs

$$B_1 = -4 \frac{j\mu_0}{2\pi} \int_0^\alpha \int_r^{r+w} \frac{\cos\theta}{\rho} \rho d\rho d\theta = -\frac{2j\mu_0}{\pi} w \sin\alpha$$



This simple computation is full of consequences

- $B_1$   $\propto$  current density (obvious)
- $B_1$   $\infty$  coil width w (less obvious)
- $B_1$  is independent of the aperture r (much less obvious)



• For a  $\cos \theta$ ,

$$B_{1} = -4 \frac{j\mu_{0}}{2\pi} \int_{0}^{\pi/2} \int_{r}^{r+w} \frac{\cos^{2}\theta}{\rho} \rho d\rho d\theta = -\frac{j\mu_{0}}{2} w$$

#### SECTOR COILS FOR DIPOLES



Multipoles of a sector coil

$$C_{n} = -2 \frac{j\mu_{0}R_{ref}^{n-1}}{2\pi} \int_{-\alpha}^{\alpha} \int_{r}^{r+w} \frac{\exp(-in\theta)}{\rho^{n}} \rho d\rho d\theta = -\frac{j\mu_{0}R_{ref}^{n-1}}{\pi} \int_{-\alpha}^{\alpha} \exp(-in\theta) d\theta \int_{r}^{r+w} \rho^{1-n} d\rho$$

for n=2 one has

$$B_2 = -\frac{j\mu_0 R_{ref}}{\pi} \sin(2\alpha) \log\left(1 + \frac{w}{r}\right)$$

and for n>2

$$B_{n} = -\frac{j\mu_{0}R_{ref}^{n-1}}{\pi} \frac{2\sin(\alpha n)}{n} \frac{(r+w)^{2-n} - r^{2-n}}{2-n}$$

- Main features of these equations
  - Multipoles n are proportional to sin ( n angle of the sector)
    - They can be made equal to zero!
  - Proportional to the inverse of sector distance to power n
    - High order multipoles are not affected by coil parts far from the centre

### Using free parameters



First allowed multipole B<sub>3</sub> (sextupole)

$$B_3 = \frac{\mu_0 j R_{ref}^2}{\pi} \frac{\sin(3\alpha)}{3} \left( \frac{1}{r} - \frac{1}{r+w} \right)$$

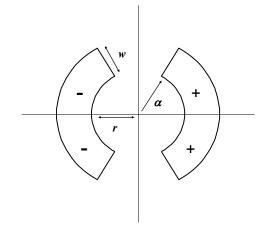
for  $\alpha = \pi/3$  (i.e. a 60° sector coil) one has  $B_3 = 0$ 

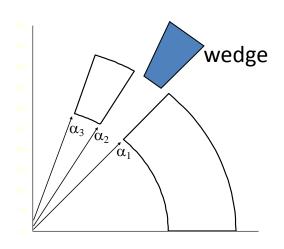
Second allowed multipole B<sub>5</sub> (decapole)

$$B_5 = \frac{\mu_0 j R_{ref}^4}{\pi} \frac{\sin(5\alpha)}{5} \left( \frac{1}{r^3} - \frac{1}{(r+w)^3} \right)$$

for  $\alpha=\pi/5$  (i.e. a 36° sector coil) or for  $\alpha=2\pi/5$  (i.e. a 72° sector coil) one has  $B_5=0$ 

• With one sector one cannot set to zero both multipoles ... but it can be done with more sectors!

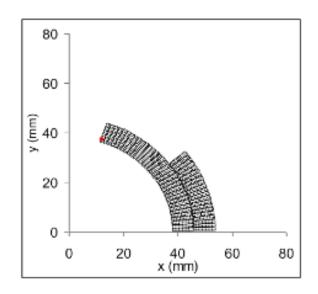




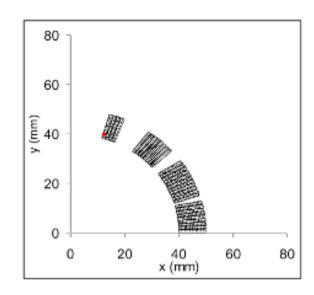
# Examples of real magnets



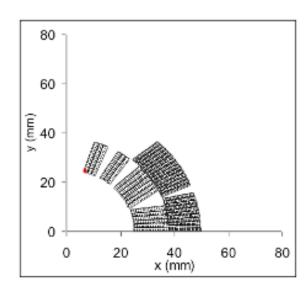
- Number of sectors is chosen based on:
  - Multipole content that can be tolerated
  - Fabrication issues



Tevatron main dipole location of the peak field



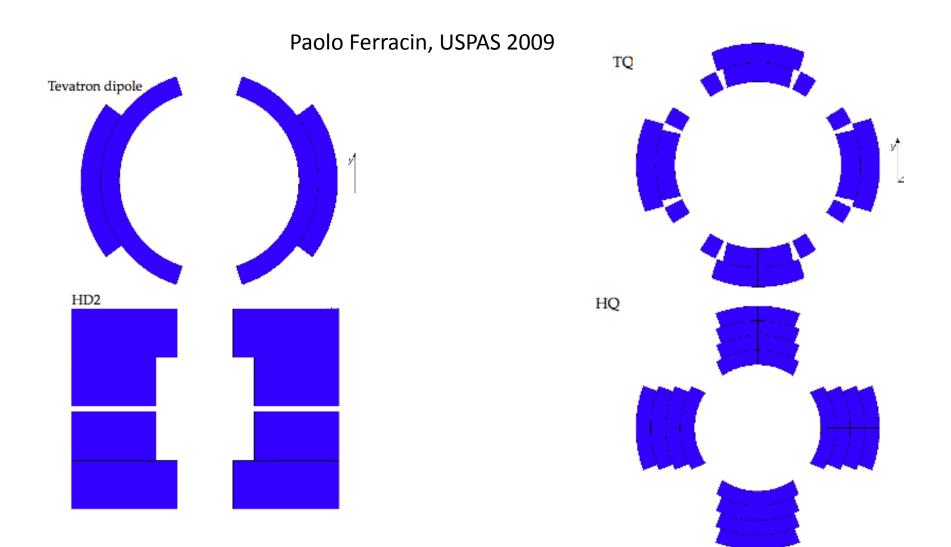
RHIC main dipole location of the peak field



LHC main dipole – location of the peak field

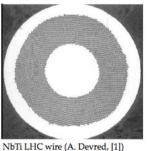
# Example geometries for real superconducting accelerator magnets

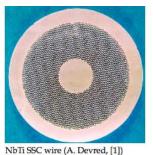




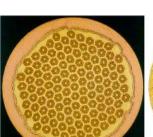
## Real superconducting magnets: Basic design / fabrication











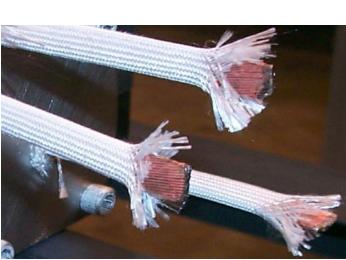
Nb<sub>∞</sub>Sn bronze-process wire

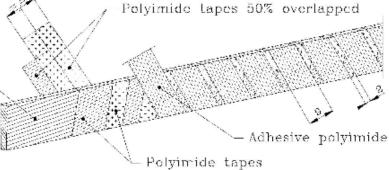
(A. Devred, [1])





Nb<sub>s</sub>Sn PIT process wire (A. Devred, [1])

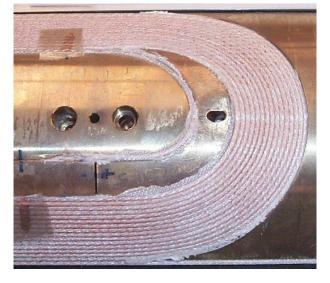




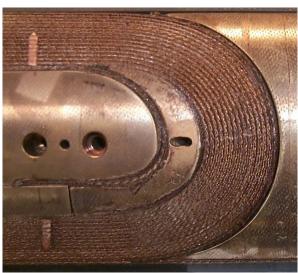
### Overview of Nb<sub>3</sub>Sn coil fabrication stages



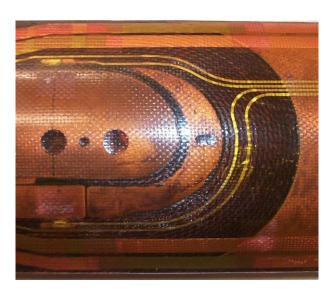
After winding



**After reaction** 

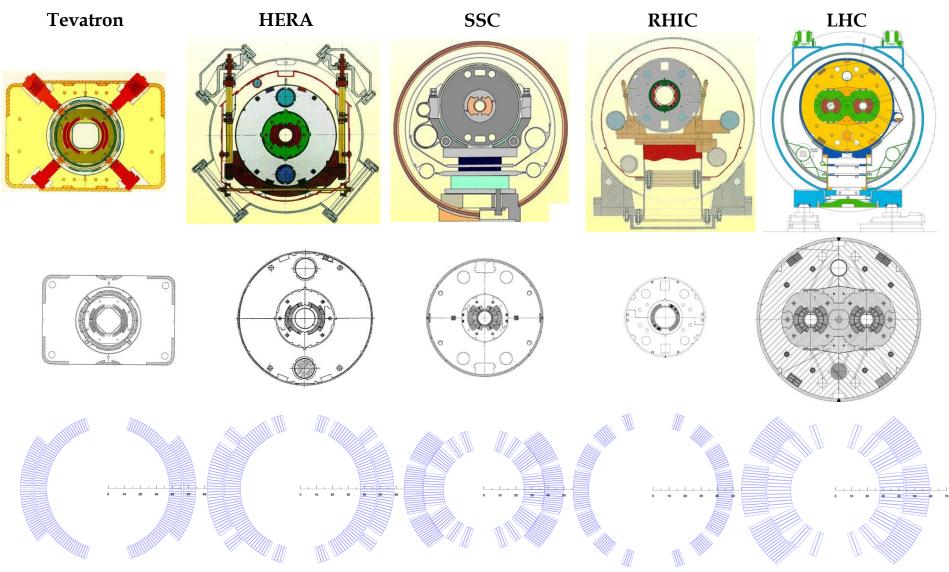


After impregnation



### Overview of accelerator dipole magnets





# Design issues

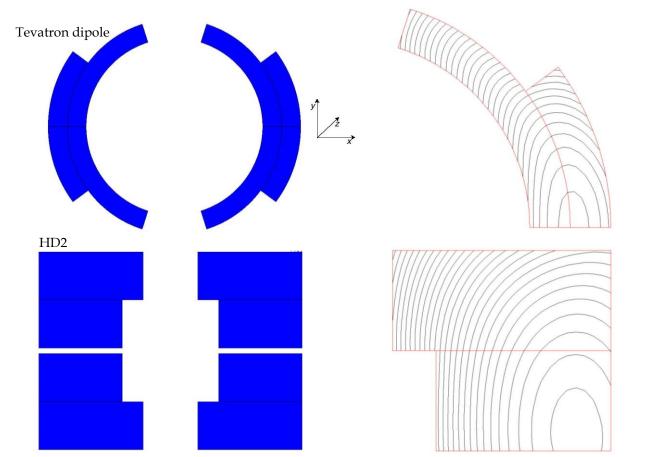


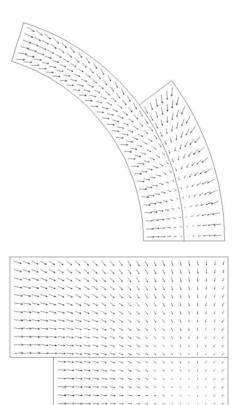
- Superconducting magnets store energy in the magnetic field
  - Results in significant mechanical stresses via Lorentz forces acting on the conductors; these forces must be controlled by structures
  - Conductor stability concerns the ability of a conductor in a magnet to withstand small thermal disturbances, e.g. conductor motion or epoxy cracking, fluxoid motion, etc.
  - The stored energy can be extracted either in a controlled manner or through sudden loss of superconductivity, e.g. via an irreversible instability – a <u>quench</u>
    - In the case of a quench, the stored energy will be converted to heat; magnet protection concerns the design of the system to appropriately distribute the heat to avoid damage to the magnet

### Lorentz force - Dipole magnets



- The Lorentz forces in a dipole magnet tend to push the coil
  - Towards the mid plane in the vertical-azimuthal direction ( $F_{\nu}$ ,  $F_{\theta}$ < 0)
  - Outwards in the radial-horizontal direction  $(F_{x_r}F_r > 0)$

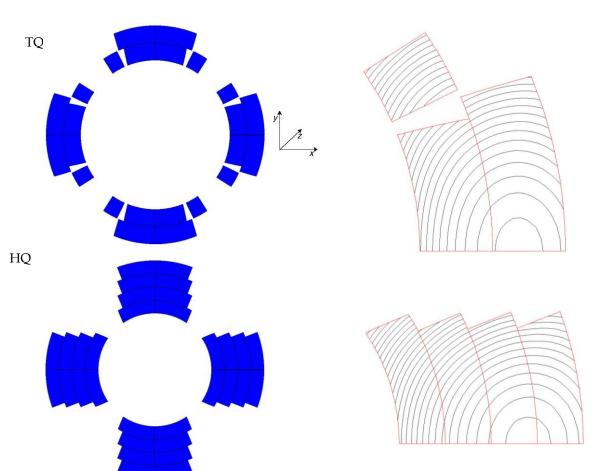


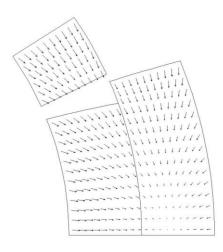


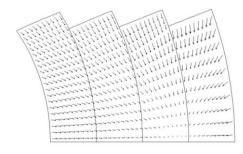
### Lorentz force - Quadrupole magnets



- The Lorentz forces in a quadrupole magnet tend to push the coil
  - Towards the mid plane in the vertical-azimuthal direction ( $F_{\nu}$ ,  $F_{\theta}$ < 0)
  - Outwards in the radial-horizontal direction  $(F_{x_r}, F_r > 0)$

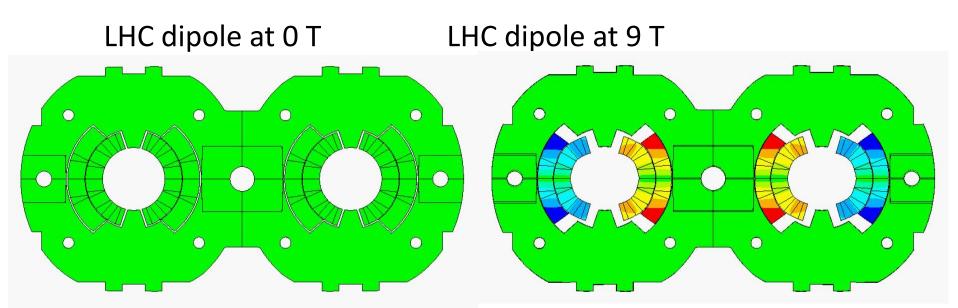






# Stress and strain Mechanical design principles





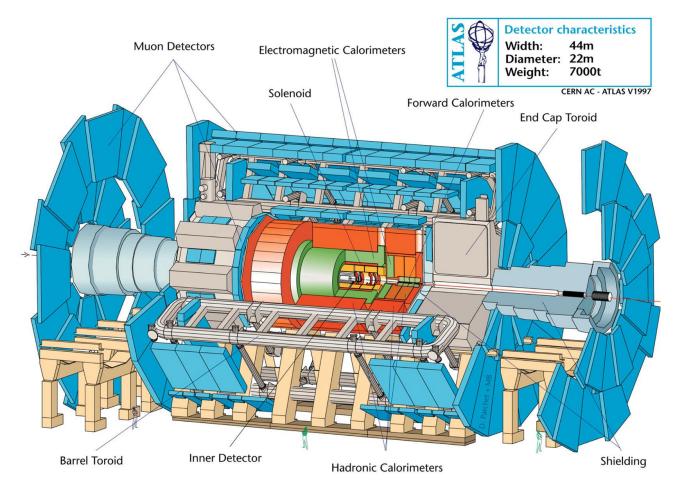
Displacement scaling = 50

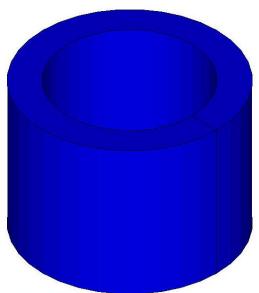
 Usually, in a dipole or quadrupole magnet, the highest stresses are reached at the mid-plane, where all the azimuthal Lorentz forces accumulate (over a small area).

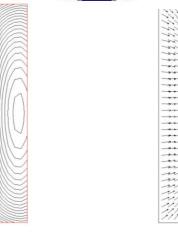
### Lorentz force - Solenoids



- The Lorentz forces in a solenoid tend to push the coil
  - Outwards in the radial-direction  $(F_r > 0)$
  - Towards the mid plane in the vertical direction  $(F_{\nu} < 0)$



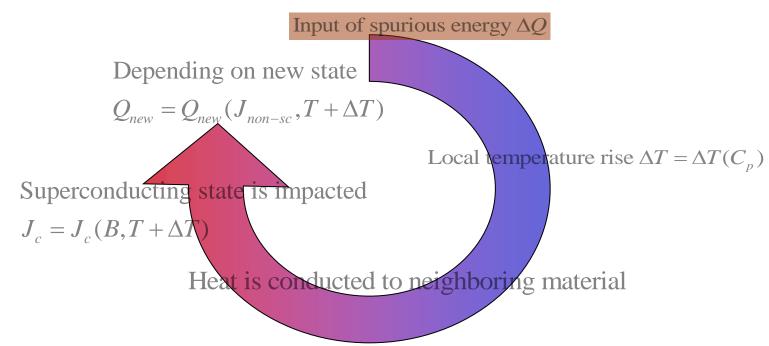




# Concept of stability



- The concept of superconductor stability concerns the interplay between the following elements:
  - The addition of a (small) thermal fluctuation local in time and space
  - The heat capacities of the neighboring materials, determining the local temperature rise
  - The thermal conductivity of the materials, dictating the effective thermal response of the system
  - The critical current dependence on temperature, impacting the current flow path
  - The current path taken by the current and any additional resistive heating sources stemming from the initial disturbance



# Calculation of the bifurcation point for superconductor instabilities



Heat Balance Equation in 1D, without coolant:  $[W/m^3]$ 

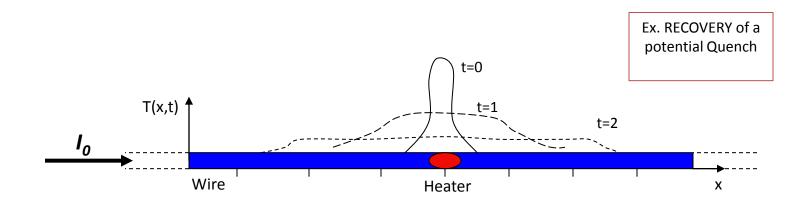
Thanks to Matteo Allesandrini, Texas Center for Superconductivity, for these calculations and slides

$$\frac{d}{dx}\left(k(T)\cdot\frac{dT}{dx}\right) + \rho(T)/J^2 + Q_{initial\_pulse} - C(T)_{volume} \cdot \frac{dT}{dt} = 0$$
Heat conduction

Joule effect

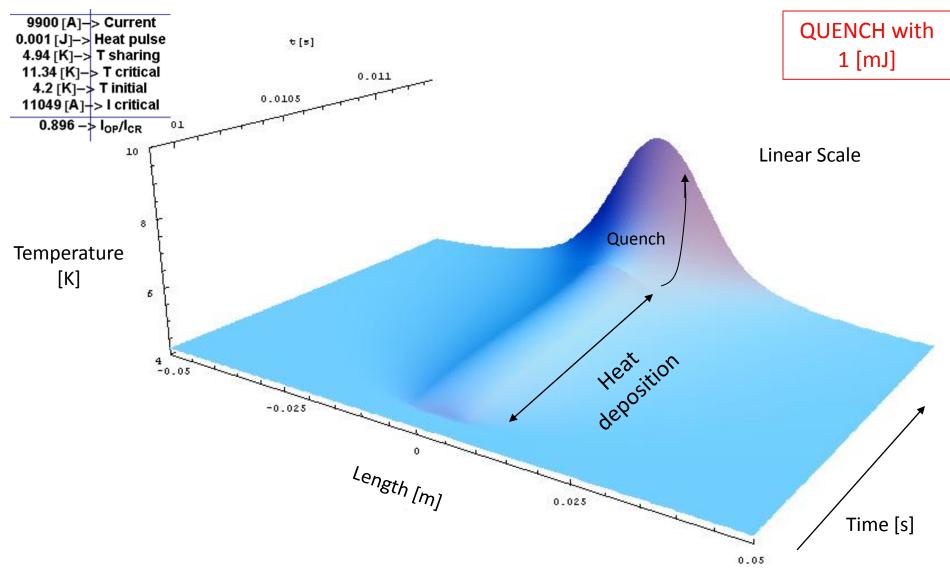
Quench trigger

Heat stored in the material



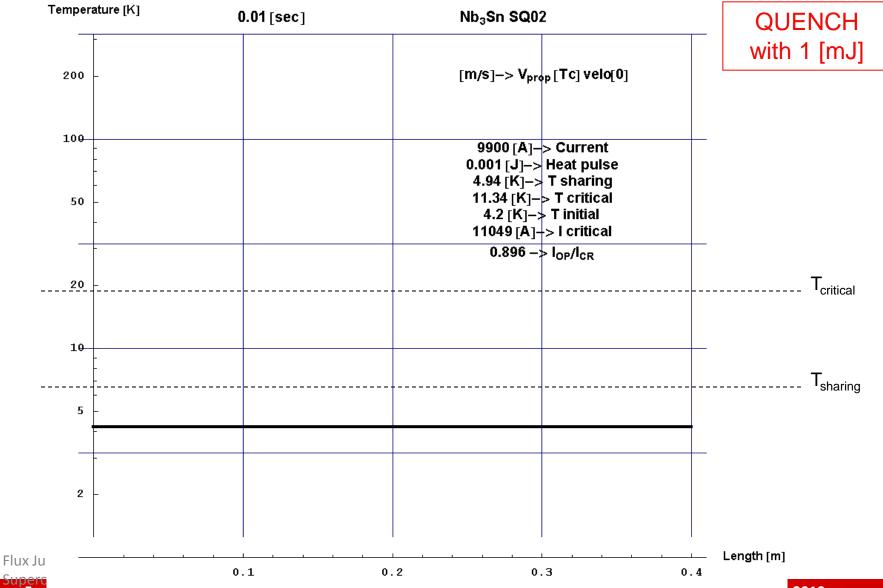
# Example of quench initiation





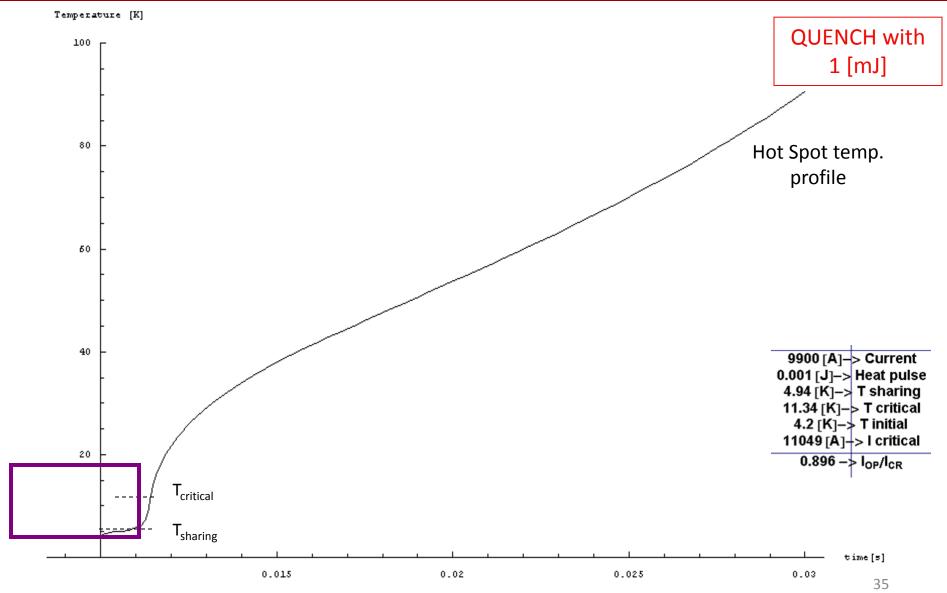
### Analysis of SQ02 – quench propagation





### Analysis of SQ02 – quench propagation

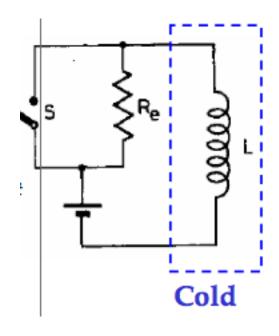


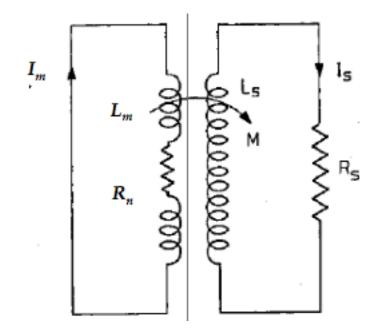


# Magnet protection



- The quench propagation aids in distributing stored energy to the rest of the magnet
- Often we accelerate the process by actively heating the magnet once a quench initiation has been detected ("Active protection")
- If possible, much of the energy is also absorbed by a dump resistor
- The energy can also be absorbed by inductive coupling to a secondary





# Permeability and field-lines



Problem 1: find the functional relationship  $\alpha 1$ =f( $\mu 1, \mu 2, \alpha 2$ ). Plot the function for  $\mu 1$ =1,  $\mu 2$ =1 and  $\mu 2$ =10

